

## FEATURES

- Wide Supply Voltage Range: 6...35V
- Wide Operating Temperature Range:  $-40^{\circ}\text{C}...+85^{\circ}\text{C}$
- Adjustable Voltage Reference: 4.5 to 10V
- Additional Current/Voltage Source
- Instrumentation Amplifier Input (Reversible Polarity)
- Operation Amplifier Input
- Adjustable Gain and Offset
- Adjustable Output Voltage Range: 0.5...4.5V, 0...5/10V, other
- Protection Against Reverse Polarity
- Output Current Limitation

## APPLICATIONS

- Industrial Process Control
- Sensor Transmitter (e.g. pressure)
- Voltage Transmitter

## GENERAL DESCRIPTION

The AM401 is a low cost monolithic voltage transmitter, designed for flexible bridge input signal conditioning. The integrated circuit is ideally suited for a wide variety of transducers with an differential output signal. It contains a high accuracy instrumentation amplifier for differential input signals, an operational amplifier output stage, and an adjustable 5 to 10V reference. In addition to these functional elements an auxiliary operational amplifier can be used as a current or voltage source. Output range and gain are adjustable by external resistors. Using the internal instrumentation amplifier the AM401 is a standard sensor transmitter with the possibility to indicate an over range signal. With the freely connectable operational amplifier input this IC can be used as an adjustable voltage-to-voltage transmitter.

## DELIVERY

- DIL16 packages (samples)
- SO16(n) packages
- Dice on 5" blue foil

## BLOCK DIAGRAM+

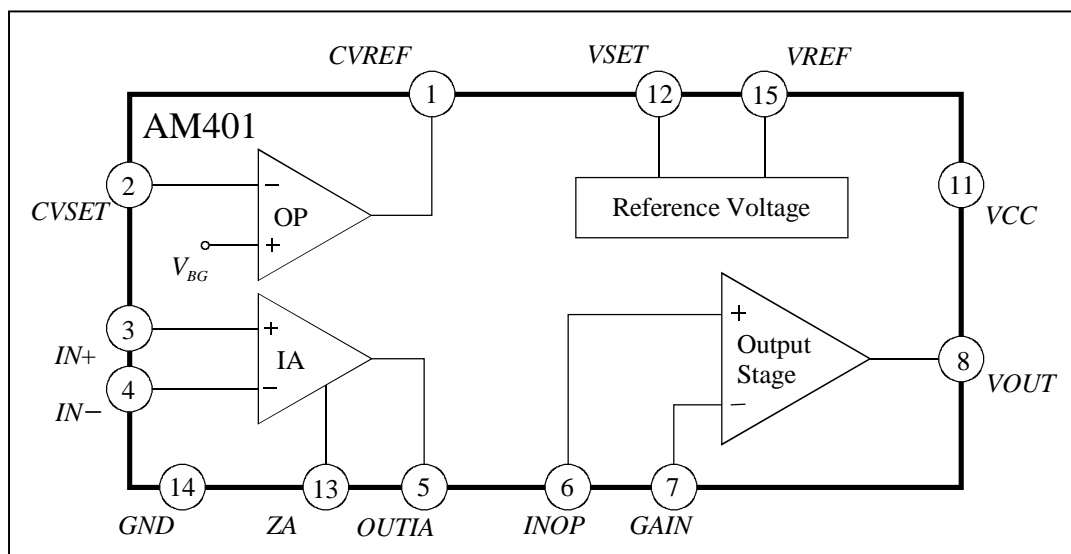


Figure 1

## ELECTRICAL SPECIFICATIONS

$T_{amb} = 25^{\circ}\text{C}$ ,  $V_{CC} = 24\text{V}$ ,  $V_{REF} = 5\text{V}$ ,  $I_{REF} = 1\text{mA}$  (unless otherwise noted)

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Voltage Range	$V_{CC}$		6		35	V
Quiescent Current	$I_{CC}$	$T_{amb} = -40\dots+85^{\circ}\text{C}$ , $I_{REF} = 0\text{mA}$			1.5	mA
<b>Temperature Specifications</b>						
Operating	$T_{amb}$		-40		85	$^{\circ}\text{C}$
Storage	$T_{st}$		-55		125	$^{\circ}\text{C}$
Junction	$T_J$				150	$^{\circ}\text{C}$
Thermal Resistance	$\Theta_{ja}$	DIL16 plastic package		70		$^{\circ}\text{C}/\text{W}$
	$\Theta_{ja}$	SO16 narrow plastic package		140		$^{\circ}\text{C}/\text{W}$
<b>Voltage Reference</b>						
Voltage	$V_{REF}$	$VSET$ not connected	4.75	5.00	5.25	V
	$V_{REF}$	$VSET = GND$ , $V_{CC} \geq 11\text{V}$	9.5	10.0	10.5	V
Trim Range	$V_{R10}$		4.5		$V_{R10}$	V
Current	$I_{REF}$		0.2		10.0	mA
$V_{REF}$ vs. Temperature	$dV_{REF}/dT$	$T_{amb} = -40\dots+85^{\circ}\text{C}$		$\pm 90$	$\pm 140$	ppm/ $^{\circ}\text{C}$
Line Regulation	$dV_{REF}/dV$	$V_{CC} = 6\text{V}\dots35\text{V}$		30	80	ppm/V
	$dV_{REF}/dV$	$V_{CC} = 6\text{V}\dots35\text{V}$ , $I_{REF} \approx 5\text{mA}$		60	150	ppm/V
Load Regulation	$dV_{REF}/dI$			0.05	0.10	%/mA
	$dV_{REF}/dI$	$I_{REF} \approx 5\text{mA}$		0.06	0.15	%/mA
Load Capacitance	$C_L$		1.9	2.2	5.0	$\mu\text{F}$
<b>Current/Voltage Source</b>						
Internal Reference	$V_{BG}$		1.20	1.27	1.35	V
$V_{BG}$ vs. Temperature	$dV_{BG}/dT$	$T_{amb} = -40\dots+85^{\circ}\text{C}$		$\pm 60$	$\pm 140$	ppm/ $^{\circ}\text{C}$
Current Source: $I_{CV} = V_{BG}/R_{EXT}$						
Adjustable Current Range	$I_{CV}$		0		10	mA
Output Voltage	$V_{CV}$	$V_{CC} < 18\text{V}$	$V_{BG}$		$V_{CC} - 5$	V
	$V_{CV}$	$V_{CC} \geq 18\text{V}$	$V_{BG}$		13	V
Voltage Source: $V_{CV} = V_{BG}(R_{EXT1} + R_{EXT2}) / R_{EXT2}$						
Adjustable Voltage Range	$V_{CV}$	$V_{CC} < 18\text{V}$	0.4		$V_{CC} - 5$	V
	$V_{CV}$	$V_{CC} \geq 18\text{V}$	0.4		13	V
Output Current	$I_{CV}$	Source			10	mA
	$I_{CV}$	Sink			-100	$\mu\text{A}$
Load Capacitance	$C_L$	Source mode	0	1	10	nF
<b>Instrumentation Amplifier</b>						
Internal Gain	$G_{IA}$		4.9	5	5.1	
Differential Input Voltage Range	$V_{IN}$		0		$\pm 400$	mV
Common Mode Input Range	$CMIR$	$V_{CC} < 9\text{V}$ , $I_{CV} < 2\text{mA}$	1.5		$V_{CC} - 3$	V
	$CMIR$	$V_{CC} \geq 9\text{V}$ , $I_{CV} < 2\text{mA}$	1.5		6.0	V
Common Mode Rejection Ratio	$CMRR$		80	90		dB
Power Supply Rejection Ratio	$PSRR$		80	90		dB
Offset Voltage	$V_{OS}$			$\pm 1.5$	$\pm 6$	mV
$V_{OS}$ vs. Temperature	$dV_{OS}/dT$			$\pm 5$		$\mu\text{V}/^{\circ}\text{C}$
Input Bias Current	$I_B$			-120	-300	nA
$I_B$ vs. Temperature	$dI_B/dT$			-0.35	-0.8	nA/ $^{\circ}\text{C}$

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
<b>Instrumentation Amplifier (cont.)</b>						
Output Voltage Range	$V_{OUTIA}$	$V_{CC} < 9V, R_{LIA} \leq 10k\Omega$	0*		$V_{CC} - 3$	V
	$V_{OUTIA}$	$V_{CC} \geq 9V, R_{LIA} \leq 10k\Omega$	0*		6	V
Minimum Output Voltage	$V_{OUTIAmin}$	without external load resistance $R_{LIA}$		5	17	mV
Load Capacitance	$C_L$				250	pF
<b>Zero Adjust Stage</b>						
Internal Gain	$G_{ZA}$			1		
Input Voltage	$V_{ZA}$	$V_{ZA} \leq V_{OUTIA} - G_{IA} V_{IN}$	0		$V_{OUTIA}$	V
Offset Voltage	$V_{OS}$			$\pm 0.5$	$\pm 2.0$	mV
$V_{OS}$ vs. Temperature	$dV_{OS}/dT$			$\pm 1.6$	$\pm 5$	$\mu V/^\circ C$
Input Bias Current	$I_B$			38	100	nA
$I_B$ vs. Temperature	$dI_B/dT$			24	75	pA/°C
<b>Voltage Output Stage</b>						
Adjustable Gain	$G_{OP}$		1			
Input Range	$I_R$	$V_{CC} < 10V$	0		$V_{CC} - 5$	V
	$I_R$	$V_{CC} \geq 10V$	0		5	V
Power Supply Rejection Ratio	$PSRR$		80	90		dB
Offset Voltage	$V_{OS}$			$\pm 0.5$	$\pm 2$	mV
$V_{OS}$ vs. Temperature	$dV_{OS}/dT$			$\pm 3$	$\pm 7$	$\mu V/^\circ C$
Input Bias Current	$I_B$			5	12	nA
$I_B$ vs. Temperature	$dI_B/dT$			3.5	10	pA/°C
Output Voltage Range	$V_{OUT}$	$V_{CC} < 18V$	0		$V_{CC} - 5$	V
	$V_{OUT}$	$V_{CC} \geq 18V$	0		13	V
Output Current Limitation	$I_{LIM}$	$V_{OUT} \geq 10V$	5	7	10	mA
Output Current	$I_{OUT}$		0		$I_{LIM}$	mA
Load Resistance	$R_L$		2			k $\Omega$
Load Capacitance	$C_L$				500	nF
<b>Protection Functions</b>						
Protection against reverse polarity		Ground vs. $V_S$ vs. $V_{OUT}$			35	V
<b>System Parameters</b>						
Nonlinearity		ideal input		0.05	0.15	%FS

\* Depending on external load resistance at output of IA ( $R_{LIA} \leq 10k\Omega \Rightarrow V_{OUTIA} < 3mV$ ); internal load resistance is  $\approx 100k\Omega$   
 Currents flowing into the IC are negative

## BOUNDARY CONDITIONS

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Sum Gain Resistors	$R_1 + R_2$	only for sink mode of CV reference	90		200	k $\Omega$
Sum Offset Resistors	$R_3 + R_4$		20		200	k $\Omega$
$V_{REF}$ Capacitance	$C_1$		1.9	2.2	5.0	$\mu F$
$V_{IA}$ Capacitance	$C_2$		10		100	nF

## FUNCTIONAL DIAGRAM

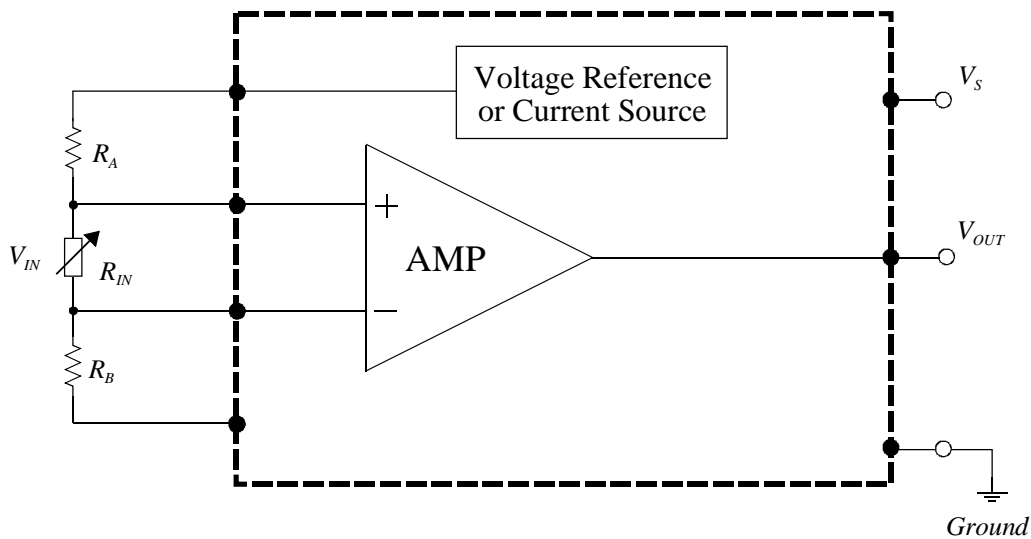


Figure 2

## FUNCTIONAL DESCRIPTION

The IC AM401 is an integrated voltage transmitter for bridge input signals. With variations of a few external components the output voltage can be adjusted over a wide range. In addition to the resistors  $R_1$  and  $R_2$  the circuitry needs only two external capacitors  $C_1$  and  $C_2$  for a basic application. Using the input of the voltage output stage the AM401 can be used for single ended input signals as well. Typical values for the external components are listed in the *Application Notes*.

Basically the AM401 consists of 4 functional blocks as they are shown in Figure 1:

1. A high accuracy *instrumentation amplifier* with an internal gain  $G_{IA}$  and the possibility to adjust the bias voltage (pin ZA) for differential input signals.
2. An *operational amplifier output stage* used for voltage transmission and as the voltage output. The output stage has an output current limitation protecting the IC.
3. An *adjustable voltage reference* ( $VSET = N.C.$  or  $VSET = GND$ ) can be used as an excitation for constant voltage sensors or as supply for other external devices.
4. An *auxiliary operational amplifier* which can be used as an additional current/voltage source or as an adjustment for the bias offset voltage (see *Application Notes*).

The transfer function of the output voltage of the instrumentation amplifier is:

$$V_{OUTIA} = G_{IA}V_{IN} + V_{ZA}$$

with the an offset voltage  $V_{ZA}$  which can be adjusted on the zero adjust pin ZA. Using a circuitry with the additional operational amplifier like shown in figure 7, this offset voltage can be calculated by

$$V_{OFFSET} = V_{BG} - \frac{R_4}{R_3}(V_{REF} - V_{BG})$$

For the entire output voltage  $V_{OUT}$  of the IC is valid

$$V_{OUT} = G_{OP} \cdot V_{INOP}$$

with the adjustable gain  $G_{OP}$

$$G_{OP} = \frac{R_1}{R_2} + 1$$

The minimum supply voltage, which has to be adjusted, can be calculated by

$$V_S \geq V_{OUTmax} + 5V$$

**Important:** Non used blocks of the IC like the auxiliary OP in Application 1 have to be connected. The two capacitors  $C_1$  and  $C_2$  have to be connected in every case as well! Even if the voltage reference is not used.

## PINOUT

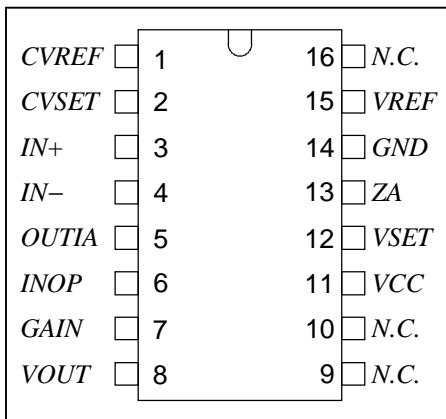


Figure 3

PIN	NAME	DESIGNATION
1	CVREF	Current/Voltage Reference
2	CVSET	Set Current/Voltage
3	IN+	Non Inverting Bridge Input
4	IN-	Inverting Bridge Input
5	OUTIA	Output Instrumentation Amplifier
6	INOP	Input Operational Amplifier
7	GAIN	Gain Adjustment
8	VOUT	Voltage Output
9	N.C.	Not Connected
10	N.C.	Not Connected
11	VCC	Supply Voltage
12	VSET	Voltage Select
13	ZA	Zero Adjust
14	GND	IC Ground
15	VREF	Reference Voltage
16	N.C.	Not Connected

## DELIVERY

The AM401 is available in version:

- 16 pin DIL packages (samples)
- SO 16 (n) packages
- Dice on 5" blue foil

## PACKAGE DIMENSIONS SO16 (n)

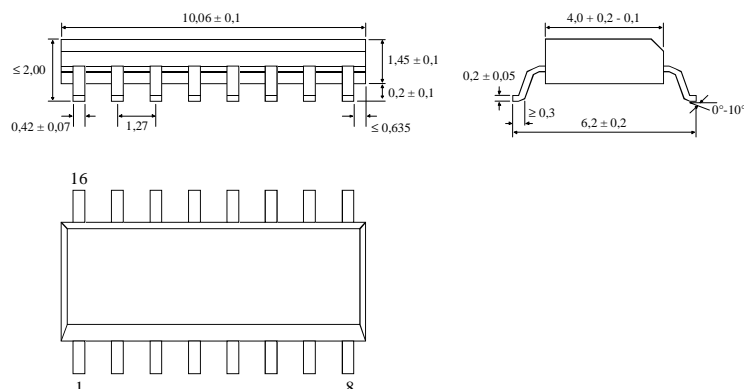
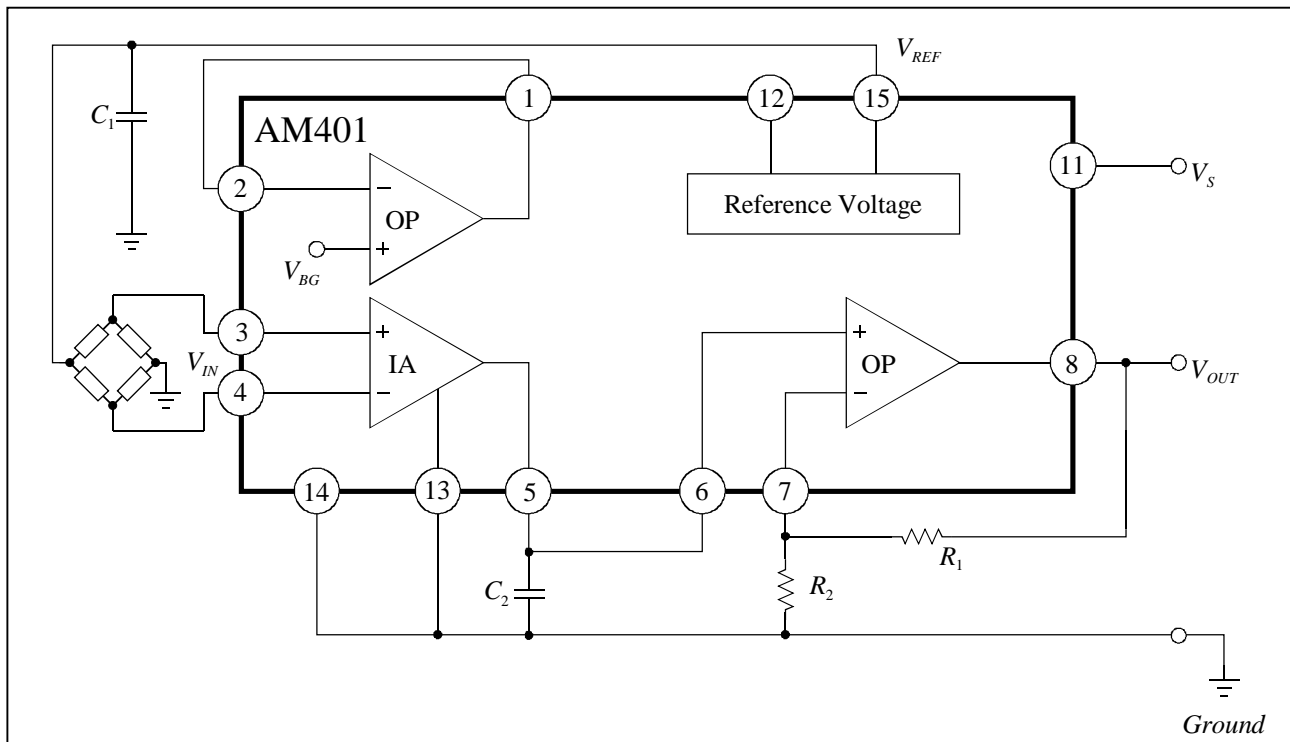


Figure 4

## APPLICATION FOR OUTPUT VOLTAGE 0...10V



**Figure 5**

Used in a 0...10V output application zero adjust pin 13 (ZA) is connected to *Ground* (Figure 5). The Gain  $G$  is adjusted by external resistors  $R_1$  and  $R_2$  and can be calculated by

$$G = G_{IA} G_{OP} = G_{IA} (1 + R_1/R_2)$$

The transfer function of the output voltage  $V_{OUT}$  becomes

$$V_{OUT} = G V_{IN}$$

With these equations the external resistors  $R_1$  and  $R_2$  can be adjusted

$$\frac{R_1}{R_2} = \frac{V_{OUT}}{G_{IA} V_{IN}} - 1$$

### Example 1: Output voltage range 0...10V

In this case the values of the external devices ( $V_{IN} = 0...50\text{mV}$ ,  $R_1/R_2 = 39$ ) are as follows

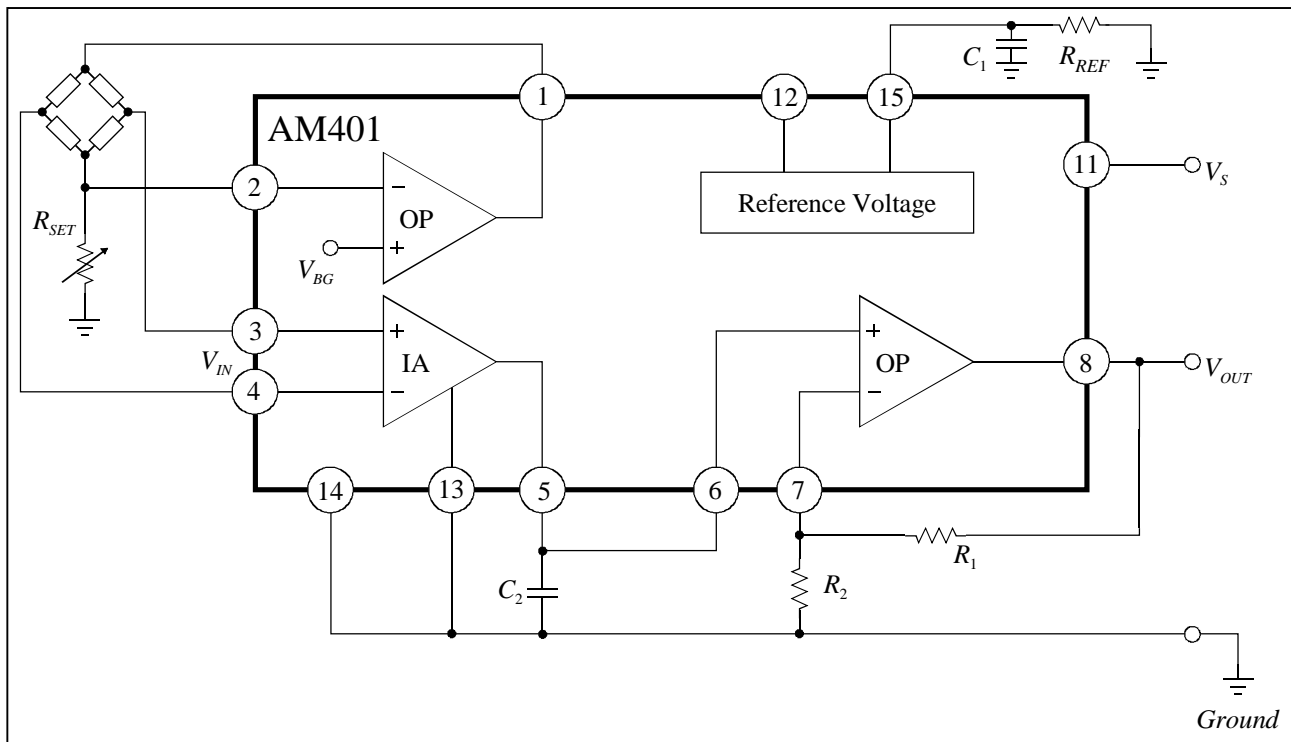
$$R_1 \approx 39\text{k}\Omega \quad R_2 \approx 1\text{k}\Omega \quad G_{IA} = 5 \quad C_1 = 2.2\mu\text{F} \quad C_2 = 10\text{nF}$$

### Example 2: Output voltage range 0...5V

In this case the values of the external devices ( $V_{IN} = 0...100\text{mV}$ ,  $R_1/R_2 = 9$ ) are as follows

$$R_1 \approx 90\text{k}\Omega \quad R_2 \approx 10\text{k}\Omega \quad G_{IA} = 5 \quad C_1 = 2.2\mu\text{F} \quad C_2 = 10\text{nF}$$

## APPLICATION FOR CURRENT DRIVEN SENSORS



**Figure 6**

In this application the auxiliary operational amplifier is used as a current source. The values of the external components are calculated for an 0...5V output. For the application the zero adjust pin 13 (ZA) is connected to *Ground* (Figure 5). The Gain  $G$  is adjusted by external resistors  $R_1$  and  $R_2$  and can be calculated by

$$G = G_{IA} G_{OP} = G_{IA} (1 + R_1/R_2)$$

The transfer function of the output voltage  $V_{OUT}$  becomes

$$V_{OUT} = G V_{IN}$$

With these equations the external resistors  $R_1$  and  $R_2$  can be adjusted

$$\frac{R_1}{R_2} = \frac{V_{OUT}}{G_{IA} V_{IN}} - 1$$

The excitation current of for the sensor can be calculated over

$$I_S = \frac{V_{BG}}{R_{SET}}$$

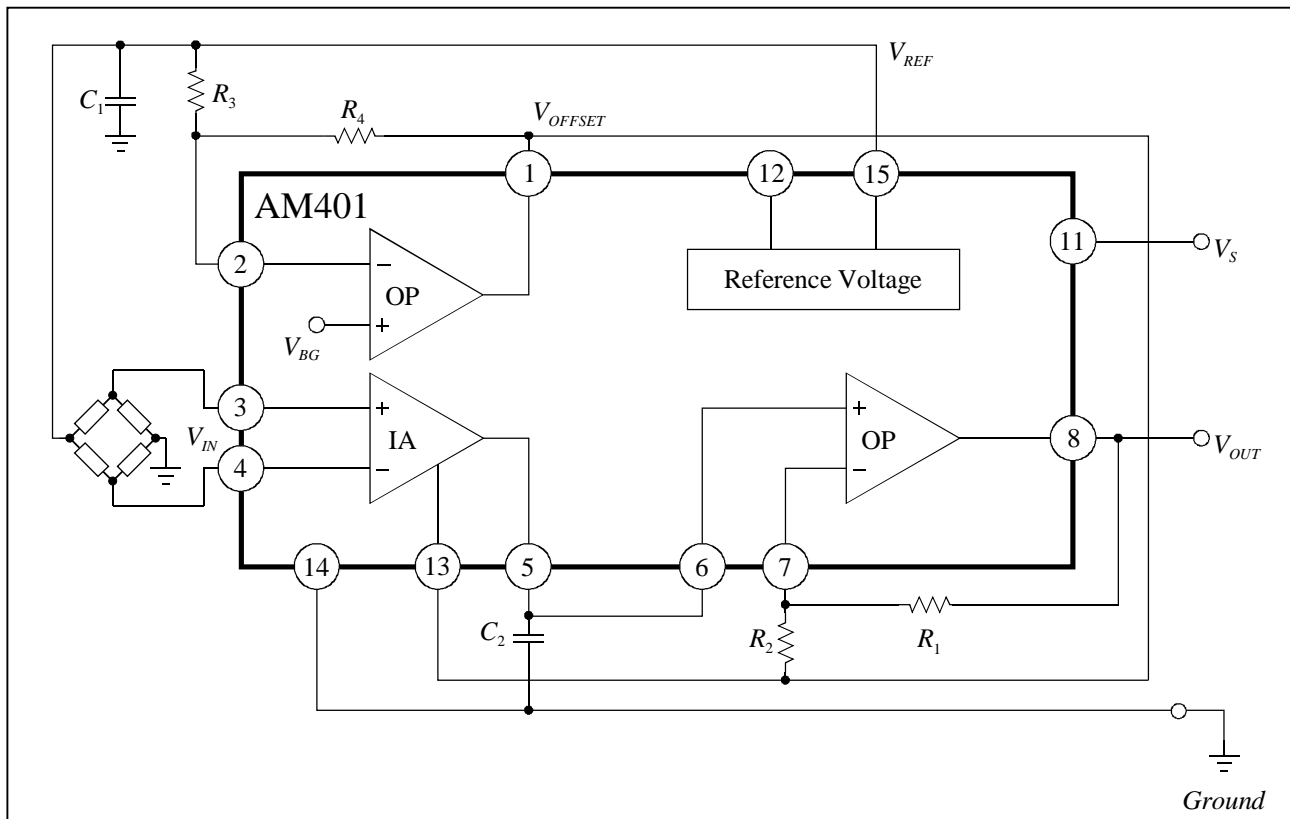
**Example 3:** Output voltage range 0...5V

The values of the external devices ( $V_{IN} = 0...100\text{mV}$ ,  $R_1/R_2 = 9$ ,  $I_S = 1.5\text{mA}$ ,  $V_{BG} = 1.27\text{V}$ ) are

$$\begin{array}{lllll} R_1 \approx 90\text{k}\Omega & R_2 \approx 10\text{k}\Omega & G_{IA} = 5 & C_1 = 2.2\mu\text{F} & C_2 = 10\text{nF} \\ R_{SET} \approx 846.7\Omega & R_{REF} \approx 22\text{k}\Omega & & & \end{array}$$



## APPLICATION FOR OUTPUT VOLTAGE 0.5...4.5V



**Figure 7**

Used in a 0.5...4.5V output application zero adjust pin 13 (ZA) is connected to an offset voltage  $V_{OFFSET}$  (Figure 6). The Gain  $G$  is adjusted by external resistors  $R_1$  and  $R_2$  and can be calculated by

$$G = G_{IA} G_{OP} = G_{IA} (1 + R_1/R_2)$$

The transfer function of the output voltage  $V_{OUT}$  becomes

$$V_{OUT} = G V_{IN} + V_{OFFSET}$$

The offset voltage can be calculated by

$$V_{OFFSET} = V_{BG} - \frac{R_4}{R_3} (V_{REF} - V_{BG}) \Rightarrow \frac{R_3}{R_4} = \frac{V_{REF} - V_{BG}}{V_{BG} - V_{OFFSET}}$$

With these equations the external resistors  $R_1$  and  $R_2$  can be adjusted

$$\frac{R_1}{R_2} = \frac{V_{OUT} - V_{OFFSET}}{G_{IA} V_{IN}} - 1$$

### Example 4: Output voltage range 0.5...4.5V

In this case the values of the external devices ( $V_{IN} = 0...250\text{mV}$ ,  $V_{REF} = 5\text{V}$ ,  $R_1/R_2 = 2.2$ ,  $R_3/R_4 = 5$ ) are as follows

$R_1 \approx 100\text{k}\Omega$	$R_2 \approx 47\text{k}\Omega$	$R_3 \approx 75\text{k}\Omega$	$R_4 \approx 15\text{k}\Omega$
$V_{OFFSET} = 0.5\text{V}$	$C_1 = 2.2\mu\text{F}$	$C_2 = 10\text{nF}$	

## TOPOLOGIE FOR THE 0...10V APPLICATION

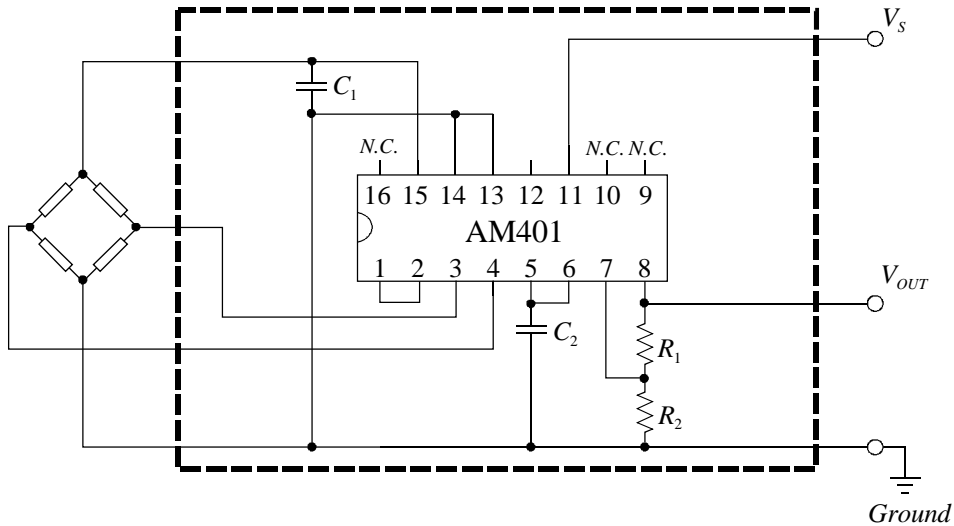


Figure 8

## TOPOLOGIE FOR THE 0.5...4.5V APPLICATION

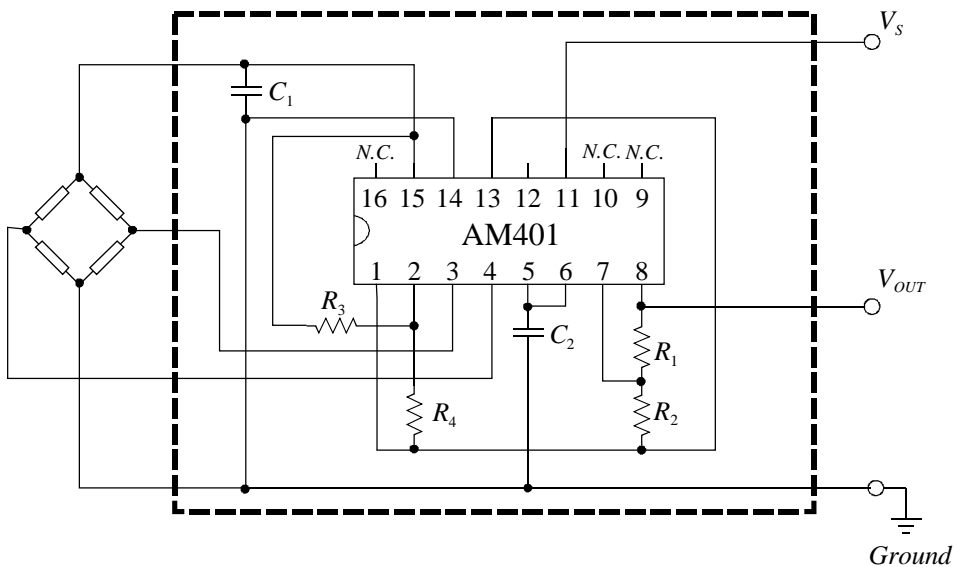


Figure 9

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